

Electromagnetics Theory

Plane-Wave Propagation

Outline:

1. Review of Waves and Phasors
2. Time-Harmonic Fields
3. Plane-Wave Propagation in Lossless Media

Introduction

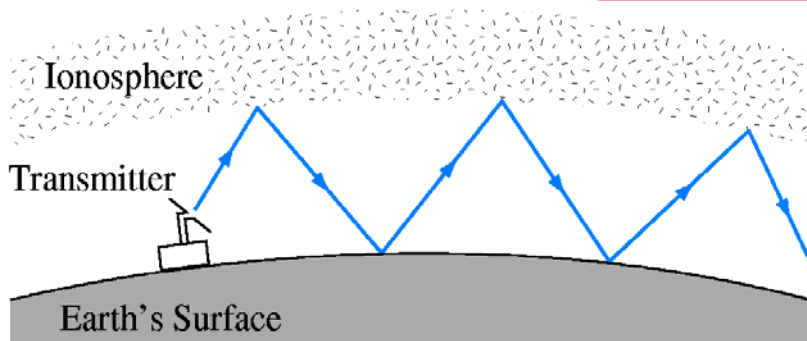
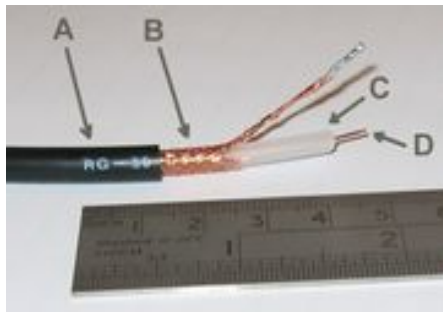
$$\mathbf{E}(t) \Leftrightarrow \mathbf{H}(t)$$

Generates EM wave

Propagate

Guided

Wireless

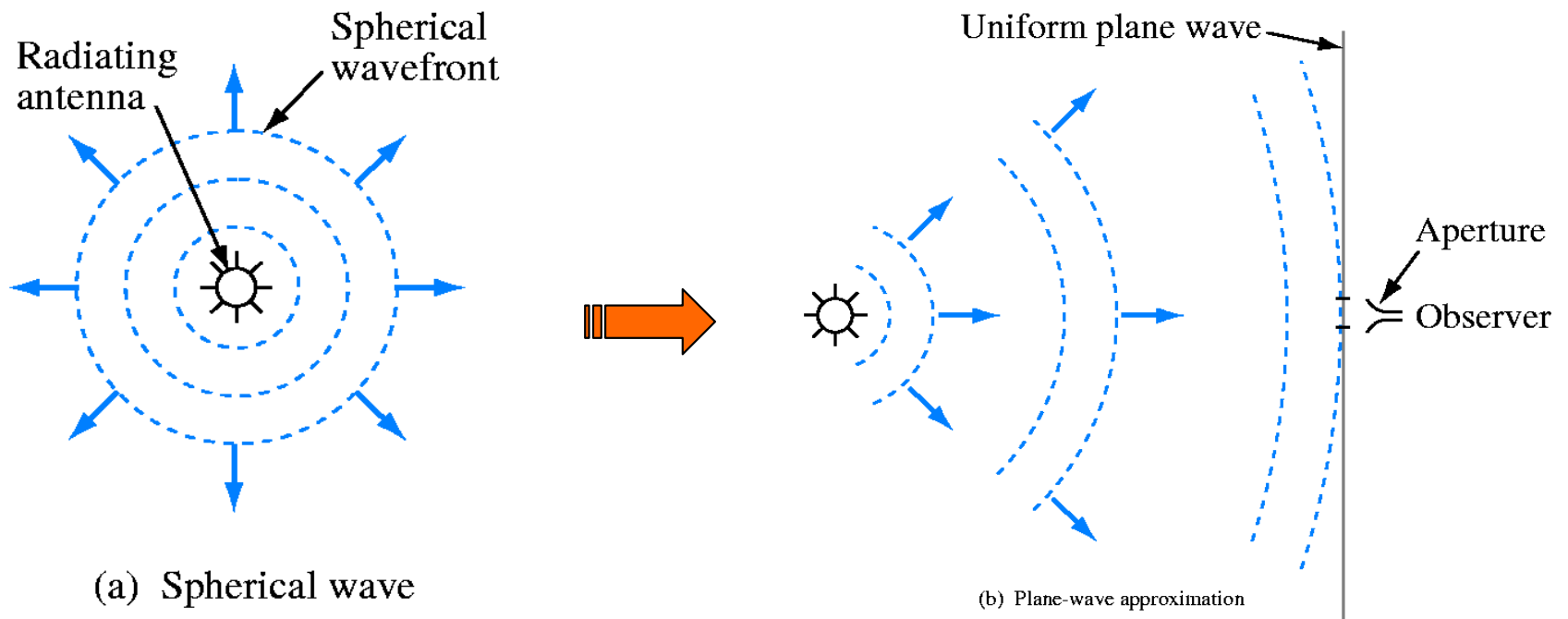


Uniform plane wave

To describe the propagation of a spherical wave



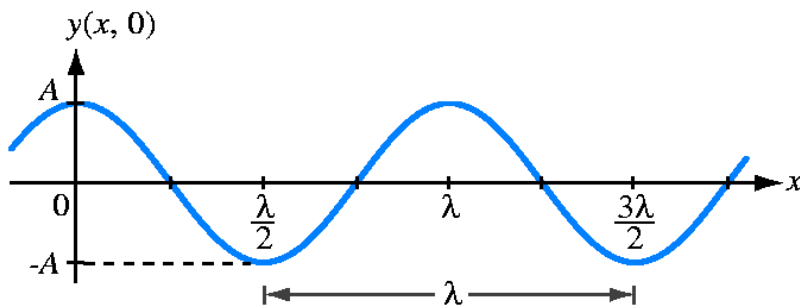
Cartesian Coordinate System



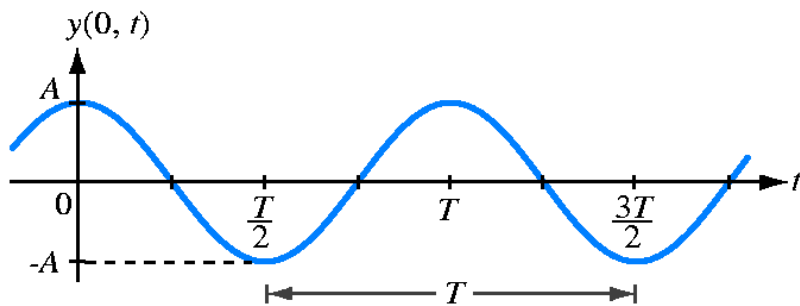
Review of Waves and Phasors

Sinusoidal Wave in a Lossless Medium

A medium is said to be **lossless** if it does not attenuate the amplitude of the wave traveling within it or on its surface.



(a) $y(x, t)$ versus x at $t = 0$



(b) $y(x, t)$ versus t at $x = 0$

$$y(x, t) = A \cos\left(\frac{2\pi t}{T} - \frac{2\pi x}{\lambda} + \phi_0\right)$$

Amplitude

Period

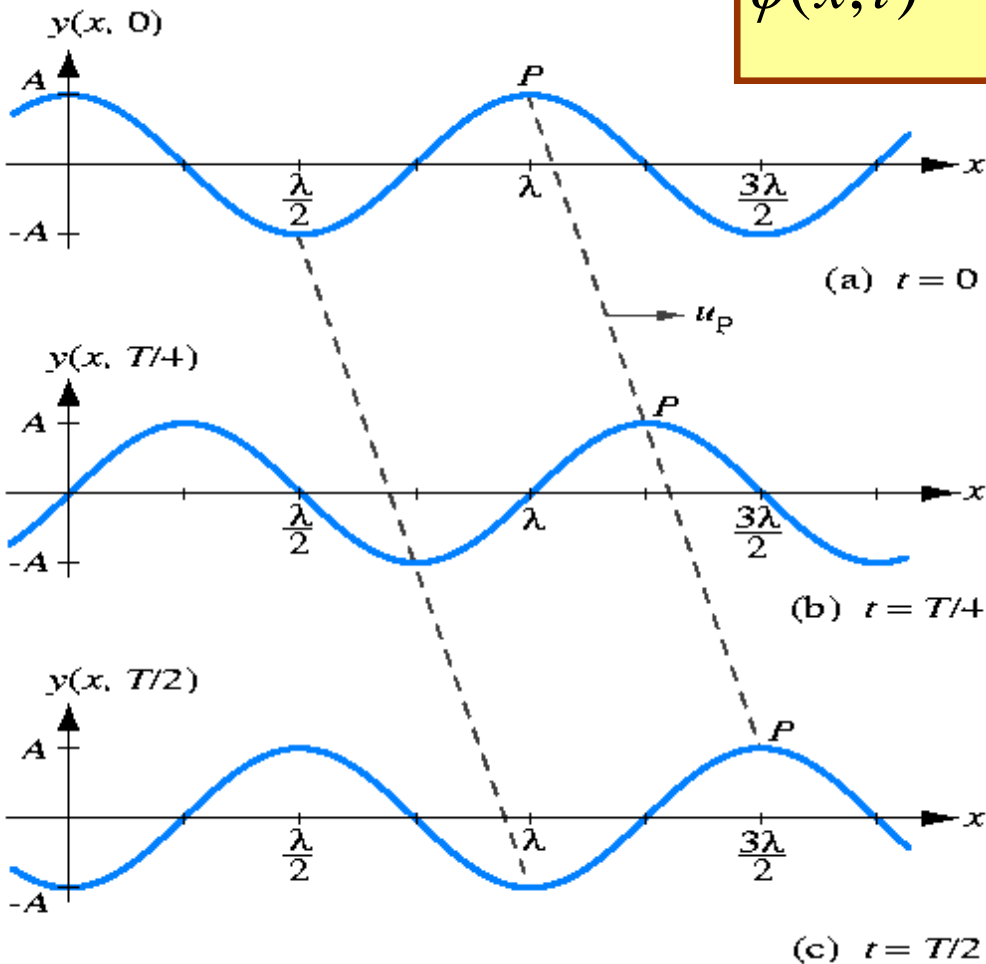
wavelength

Reference phase

$$\phi(x, t) = \left(\frac{2\pi t}{T} - \frac{2\pi x}{\lambda} + \phi_0\right) \quad \text{Phase}$$

Phase velocity

Phase velocity (u_p)



$$\phi(x, t) = \frac{2\pi t}{T} - \frac{2\pi x}{\lambda} = 2n\pi, \quad n = 0, 1, 2, \dots$$

$$\frac{d}{dt} \left(\frac{2\pi t}{T} - \frac{2\pi x}{\lambda} \right) = 0$$

$$\frac{2\pi}{T} - \frac{2\pi}{\lambda} \frac{dx}{dt} = 0$$

$$u_p = \frac{\lambda}{T} \quad (\text{m/s})$$

Phase velocity (cont'ed)

Phase velocity is the velocity of the wave pattern.

Direction of propagation

If one of the signs is +ve and the other is -ve \rightarrow wave in + x-direction

If both signs are +ve or both are -ve \rightarrow wave in - x-direction

ϕ_0 \rightarrow has no influence on either the speed or the direction of the wave propagation

$$\phi(x, t) = \left(\frac{2\pi t}{T} - \frac{2\pi x}{\lambda} + \phi_0 \right)$$

Phase velocity, phase constant

$$y(x, t) = A \cos\left(2\pi f t - \frac{2\pi}{\lambda} x\right)$$
$$= A \cos(\omega t - \beta x)$$

$$\omega = 2\pi f \quad (\text{rad/s})$$

*Angular
velocity*

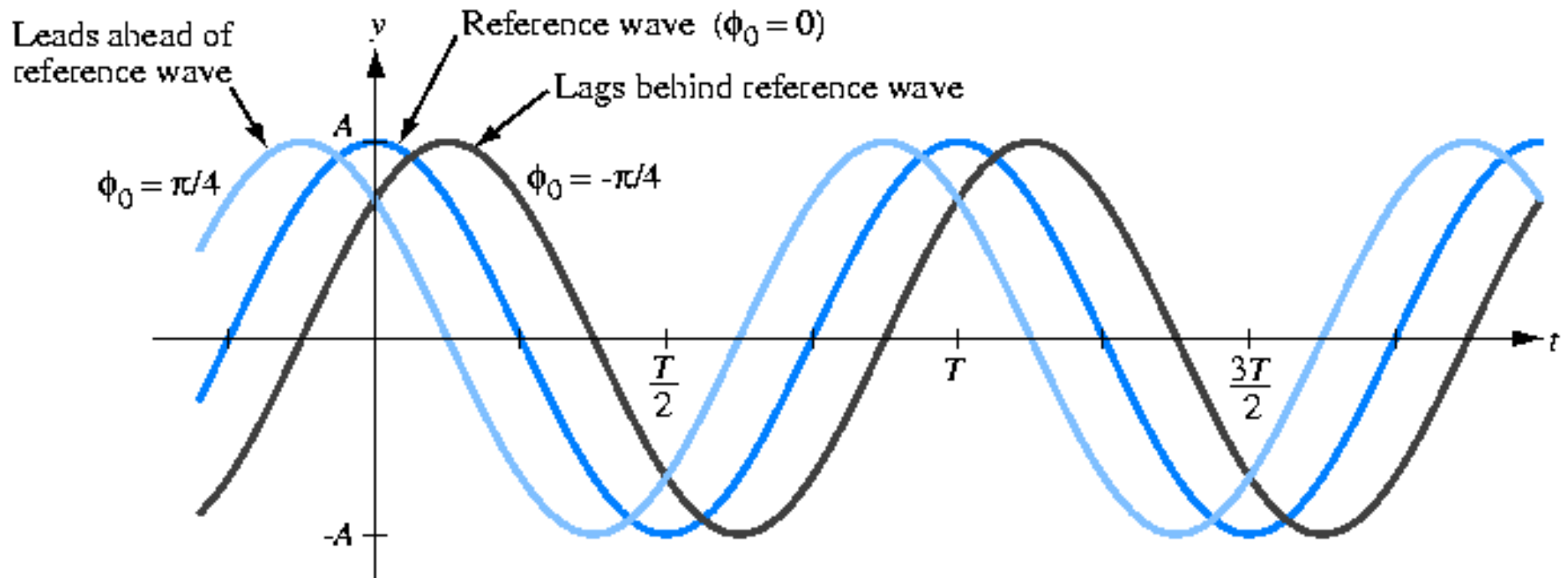
*Phase constant/
wavenumber*

$$\beta = \frac{2\pi}{\lambda} \quad (\text{rad/m})$$

u_p in terms of ω and β \longrightarrow

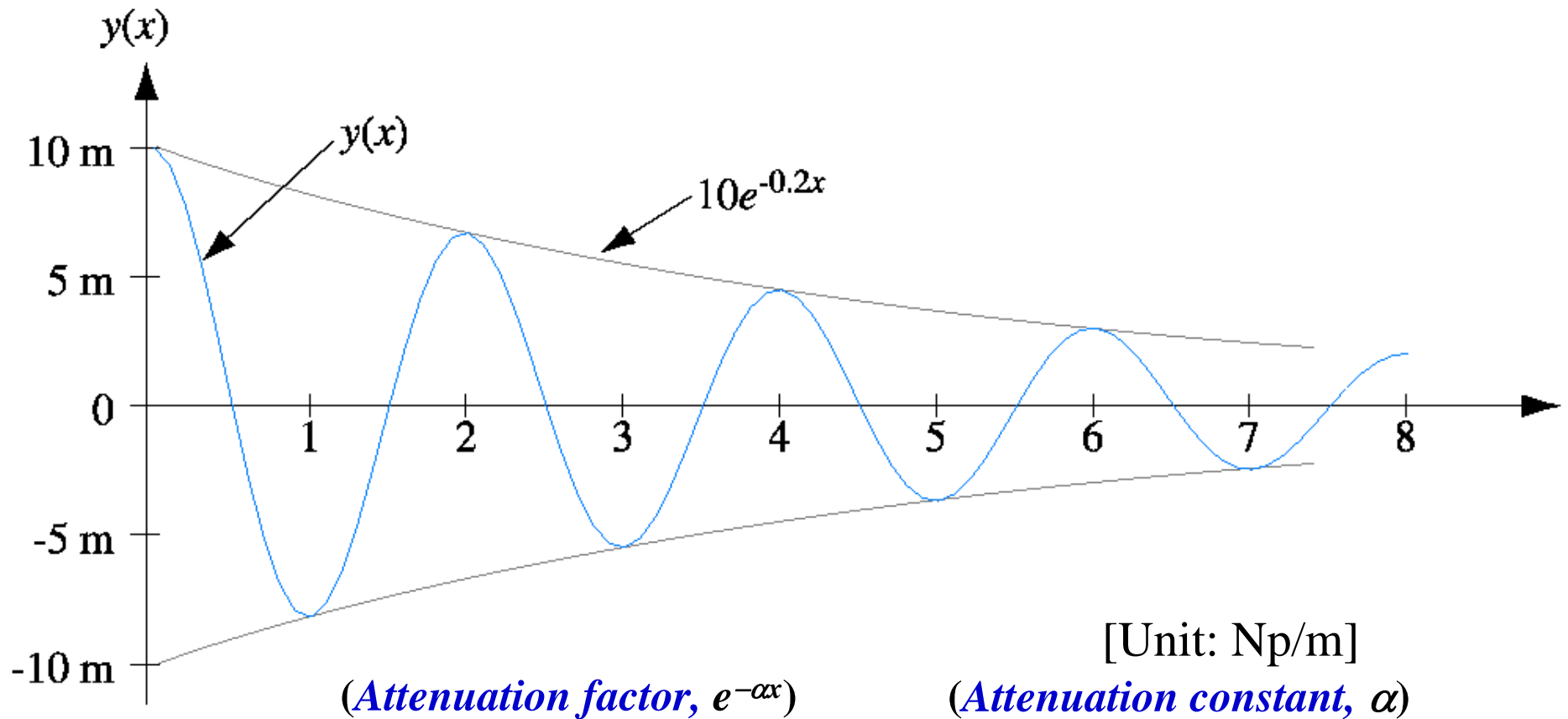
$$u_p = \frac{\omega}{\beta} = f\lambda$$

The role of reference phase, ϕ_0



$$y(x, t) = A \cos(\omega t - \beta x + \phi_0)$$

Sinusoidal Wave in a Lossy Medium



$$y(x, t) = A e^{-\alpha x} \cos(\omega t - \beta x + \phi_0)$$

Example 6-1: Sound wave in water

An acoustic wave traveling in the x -direction in a fluid is characterized by a differential pressure $p(x,t)$. The unit for pressure is N/m^2 . Find the expression for $p(x,t)$ for a sinusoidal sound wave traveling in the positive x -direction in water. Given:

$$f = 1 \text{ kHz}$$

$$u_p = 1.5 \text{ km/s}$$

$$\text{Amplitude} = 20 \text{ N/m}^2$$

$$p(x,t) \text{ max at } t = 0, x = 0.25 \text{ m.}$$

Treat water as a lossless medium

$$p(x,t) = 20 \cos\left(2\pi \times 10^3 t - \frac{4\pi}{3} x + \frac{\pi}{3}\right)$$

Example 6-2: Power Loss

A laser beam of light propagating through the atmosphere is characterized by an electric field intensity given by:

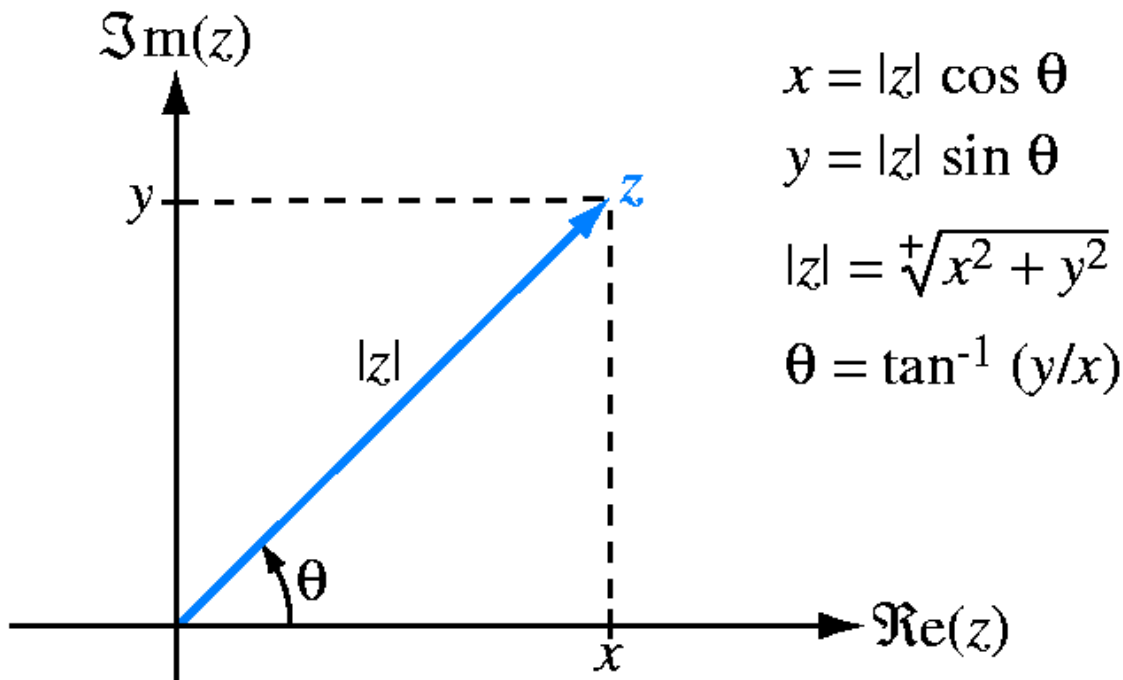
where x is the distance from the source in meters. The attenuation is due to absorption by atmospheric gases.

Determine:

- a) The direction of wave travel
- b) The wave velocity
- c) The wave amplitude at a distance of 300m

Review of Complex Numbers

Complex number



Real part Imaginary part

$$z = x + jy$$

$$z = |z| e^{j\theta} = |z| \angle \theta$$

$$e^{j\theta} = \cos \theta + j \sin \theta$$

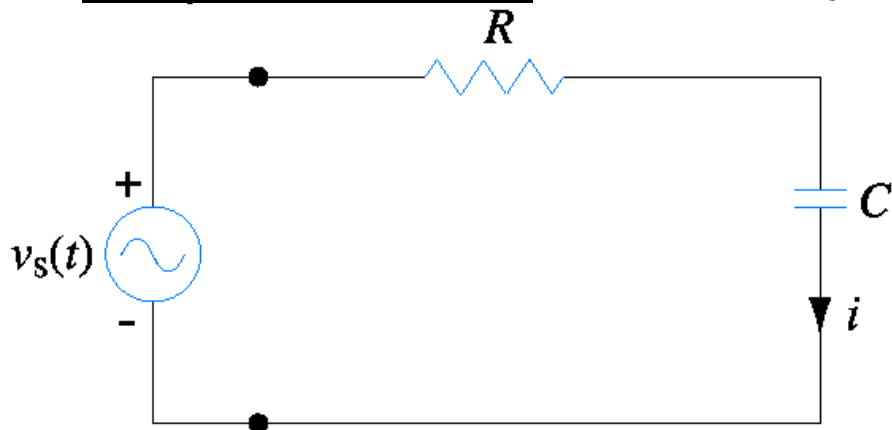
Review of Phasors

What is *phasor analysis* ?

↪ Is a mathematical tools

↪ Solving linear systems with periodic **time function**

Simple RC circuit



Objective: Obtain an expression for current $i(t)$

$$v_s(t) = V_o \sin(\omega t + \phi_0)$$

$$i(t) = ??$$

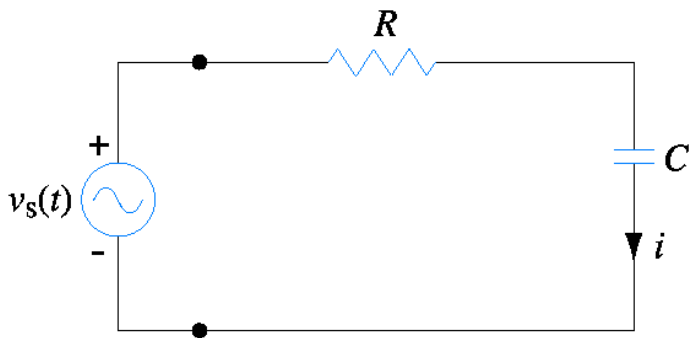
Review of Phasors (cont'ed)

Step 1: Adopt a cosine reference

$$v_s(t) = V_o \sin(\omega t + \phi_0)$$



$$v_s(t) = V_o \cos\left(\omega t + \phi_0 - \frac{\pi}{2}\right)$$



Apply KVL

$$Ri(t) + \frac{1}{C} \int i(t) dt = v_s(t) \quad (\text{time domain})$$

Review of Phasors (cont'ed)

Step 2: Express time-dependent variables as phasors

$$z(t) = \text{Re}[\tilde{Z}e^{j\omega t}] \quad \leftarrow$$

Time-independent
phasors

$$v_s(t) = V_0 \cos(\omega t + \phi_0 - \frac{\pi}{2})$$

$$\begin{aligned} v_s(t) &= \text{Re}\left[V_0 e^{(j\omega t + \phi_0 - \frac{\pi}{2})} \right] \\ &= \text{Re}\left[V_0 e^{(\phi_0 - \frac{\pi}{2})} e^{j\omega t} \right] \\ &= \text{Re}\left[\tilde{V}_s e^{j\omega t} \right] \end{aligned}$$

Recall:

$$Ri(t) + \frac{1}{C} \int i(t) dt = v_s(t) \quad (\text{time domain})$$

★ $v_s(t)$ is now represented
with time-independent
phasors

What next?

Review of Phasors (cont'ed)

$$i(t) \Rightarrow i(t) = \text{Re}(\tilde{I}e^{j\omega t})$$

Recall:

$$Ri(t) + \frac{1}{C} \int i(t) dt = v_s(t) \quad (\text{time domain})$$

Solution:

differentiation

$$\frac{di}{dt} = \text{Re}[j\omega\tilde{I}e^{j\omega t}]$$

integration

$$\int i dt = \text{Re}\left[\frac{\tilde{I}}{j\omega} e^{j\omega t}\right]$$

Review of Phasors (cont'ed)

Step 3: Recast the differential/integral form equation in phasor form

$$Ri(t) + \frac{1}{C} \int i(t) dt = v_s(t) \quad (\text{time domain})$$



$$R(\text{Re}[\tilde{I}e^{j\omega t}]) + \frac{1}{C} (\text{Re}[\frac{\tilde{I}}{j\omega} e^{j\omega t}]) = \text{Re}[\tilde{V}_s e^{j\omega t}]$$



$$\tilde{I} (R + \frac{1}{j\omega C}) = \tilde{V}_s \quad (\text{phasor domain})$$

Review of Phasors (cont'ed)

Step 4: Solve the phasor-domain equation

$$\tilde{I} = \frac{\tilde{V}_s}{R + 1/(j\omega C)} \quad \Rightarrow \quad \tilde{I} = \frac{V_o \omega C}{\sqrt{1 + \omega^2 R^2 C^2}} e^{j(\phi_0 - \phi_1)}$$

Step 5: Find the instantaneous value

Objective: Obtain $i(t)$

$$\begin{aligned} i(t) &= \text{Re}[\tilde{I} e^{j\omega t}] \\ &= \text{Re}\left[\frac{V_o \omega C}{\sqrt{1 + \omega^2 R^2 C^2}} e^{j(\phi_0 - \phi_1)} e^{j\omega t}\right] \\ &= \frac{V_o \omega C}{\sqrt{1 + \omega^2 R^2 C^2}} \cos(\omega t + \phi_o - \phi_1) \end{aligned}$$

Time-domain vs phasor-domain

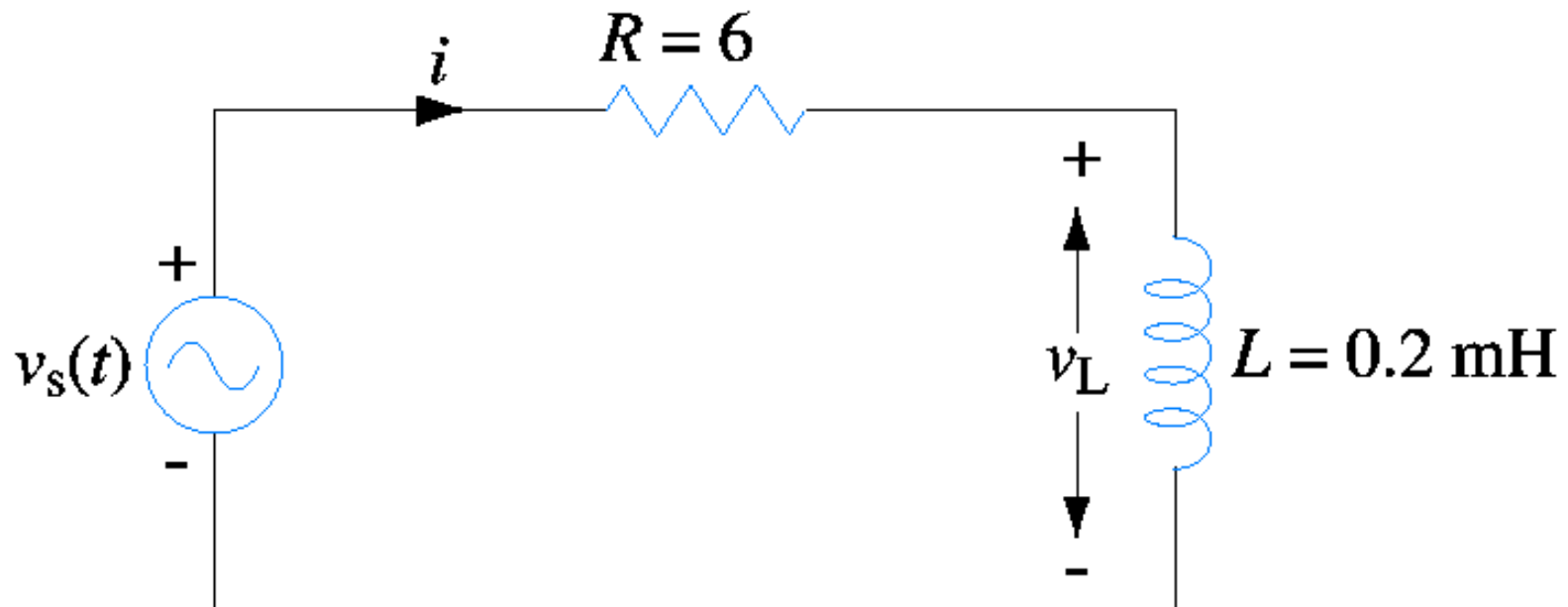
$z(t)$		\tilde{Z}
$A \cos \omega t$	\leftrightarrow	A
$A \cos(\omega t + \phi_0)$	\leftrightarrow	$A e^{j\phi_0}$
$A \cos(\omega t + \beta x + \phi_0)$	\leftrightarrow	$A e^{j(\beta x + \phi_0)}$
$A e^{-\alpha x} \cos(\omega t + \beta x + \phi_0)$	\leftrightarrow	$A e^{-\alpha x} e^{j(\beta x + \phi_0)}$
$A \sin \omega t$	\leftrightarrow	$A e^{-j\pi/2}$
$A \sin(\omega t + \phi_0)$	\leftrightarrow	$A e^{j(\phi_0 - \pi/2)}$
$\frac{d}{dt}(z_1(t))$	\leftrightarrow	$j\omega \tilde{Z}_1$
$\frac{d}{dt}[A \cos(\omega t + \phi_0)]$	\leftrightarrow	$j\omega A e^{j\phi_0}$
$\int z_1(t) dt$	\leftrightarrow	$\frac{1}{j\omega} \tilde{Z}_1$
$\int A \sin(\omega t + \phi_0) dt$	\leftrightarrow	$\frac{1}{j\omega} A e^{j(\phi_0 - \pi/2)}$

Example 6-4: *RL* Circuit

The voltage source of the circuit shown below is given by:

$$v_s(t) = 5 \sin(4 \times 10^4 t - 30^\circ)$$

Obtain an expression for the voltage across the inductor.



Time Harmonic Fields

In the *time-varying case*, the electric and magnetic fields (\mathbf{E} , \mathbf{D} , \mathbf{B} and \mathbf{H}) and source (ρ_v , \mathbf{J}) are each, in general, a function of the spatial coordinates (x, y, z) and the time variable t .

$$\mathbf{E}(x, y, z; t) = \text{Re}[\tilde{\mathbf{E}}(x, y, z)e^{j\omega t}]$$

Maxwell's equations
in phasor domain



$$\nabla \cdot \tilde{\mathbf{E}} = \tilde{\rho}_v / \epsilon$$

$$\nabla \times \tilde{\mathbf{E}} = -j\omega\mu\tilde{\mathbf{H}}$$

$$\nabla \cdot \tilde{\mathbf{H}} = 0$$

$$\nabla \times \tilde{\mathbf{H}} = \tilde{\mathbf{J}} + j\omega\epsilon\tilde{\mathbf{E}}$$

Complex Permittivity

$$\nabla \times \tilde{\mathbf{H}} = \tilde{\mathbf{J}} + j\omega\epsilon\tilde{\mathbf{E}} \quad ; \mathbf{J} = \sigma\mathbf{E}$$

$$= (\sigma + j\omega\epsilon)\tilde{\mathbf{E}}$$

$$= j\omega\left(\epsilon - j\frac{\sigma}{\omega}\right)\tilde{\mathbf{E}}$$

Complex permittivity, ϵ_c



For a lossless medium, $\sigma = 0$, $\epsilon_c = \epsilon$

Wave equations for a charge-free medium

Charge free $\equiv \rho_v = 0$

Maxwell's equations
for charge a free
medium

$$\begin{aligned}\nabla \cdot \tilde{\mathbf{E}} &= 0 \\ \nabla \times \tilde{\mathbf{E}} &= -j\omega\mu\tilde{\mathbf{H}} \\ \nabla \cdot \tilde{\mathbf{H}} &= 0 \\ \nabla \times \tilde{\mathbf{H}} &= \tilde{\mathbf{J}} + j\omega\epsilon_c\tilde{\mathbf{E}}\end{aligned}$$

Homogenous wave
equation for $\tilde{\mathbf{E}}$

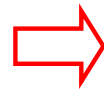
$$\nabla^2 \tilde{\mathbf{E}} + \omega^2 \mu \epsilon_c \tilde{\mathbf{E}} = 0$$

Propagation constant : $\gamma^2 = -\omega^2 \mu \epsilon_c$

Wave equations for **E**, **H**

Wave equations for **E**

$$\nabla^2 \tilde{\mathbf{E}} + \omega^2 \mu \epsilon_c \tilde{\mathbf{E}} = 0$$



$$\nabla^2 \tilde{\mathbf{E}} - \gamma^2 \tilde{\mathbf{E}} = 0$$

Wave equations for **H**

$$\nabla^2 \tilde{\mathbf{H}} + \omega^2 \mu \epsilon_c \tilde{\mathbf{H}} = 0$$



$$\nabla^2 \tilde{\mathbf{H}} - \gamma^2 \tilde{\mathbf{H}} = 0$$

Plane-Wave Propagation in Lossless Media

Lossless media \equiv *nonconducting* medium ($\sigma=0$)

$$\varepsilon_c = \varepsilon - j \frac{\sigma}{\omega}$$

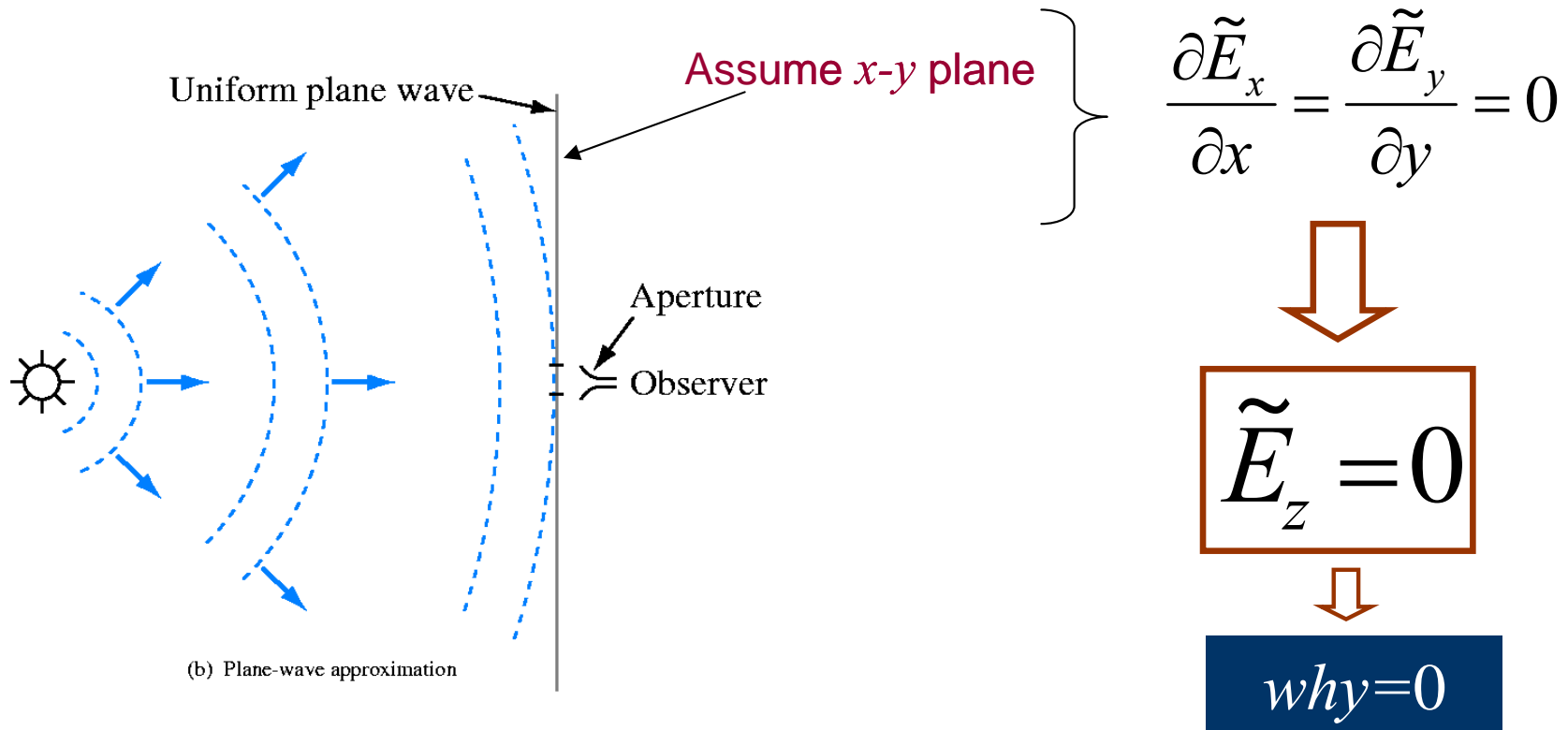
$$\gamma^2 = -\omega^2 \mu \varepsilon$$

Wavenumber k (lossless)

$$k = \omega \sqrt{\mu \varepsilon}$$

Uniform Plane Waves

A *uniform plane wave* is characterized by *electric* and *magnetic* fields that have *uniform* properties at all points across an infinite plane.



Uniform Plane Waves (cont'ed)

why=0

$$\nabla \times \tilde{\mathbf{H}} = j\omega\epsilon\tilde{\mathbf{E}}$$

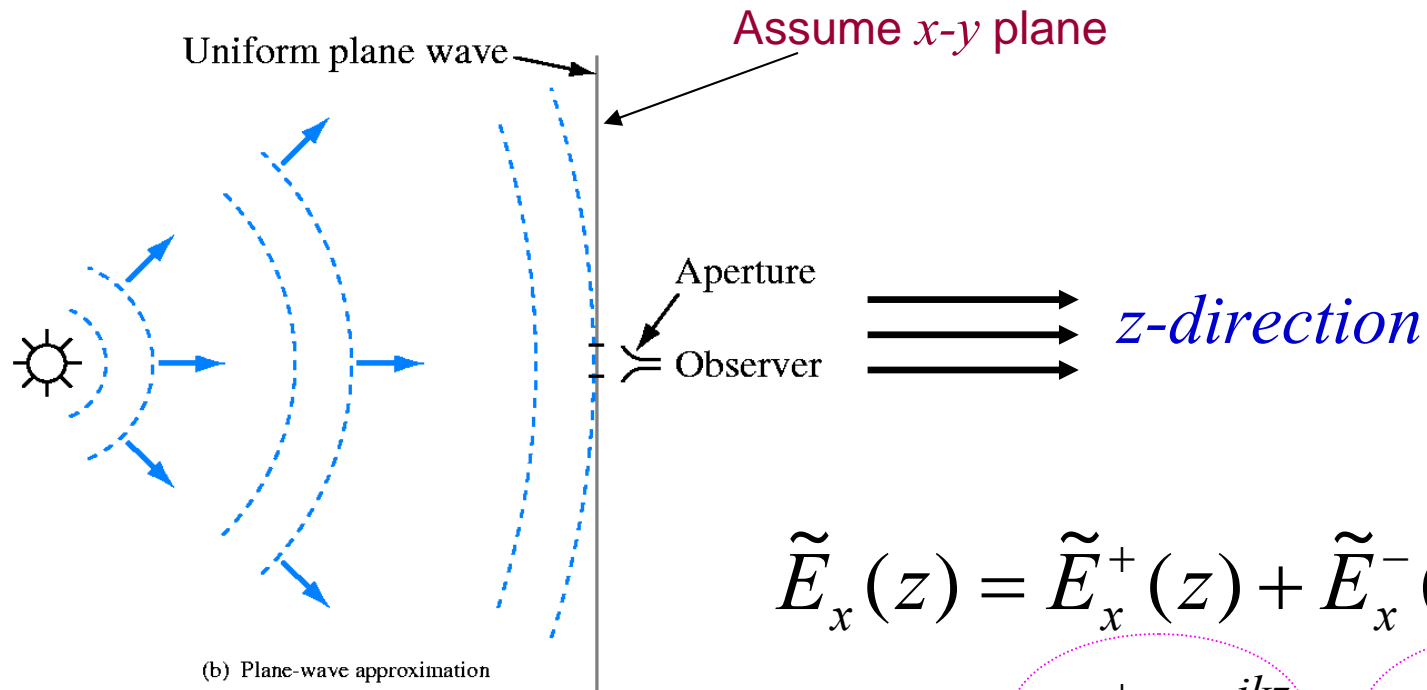
Taking the *Curl* on
the *z*-component

$$\hat{\mathbf{z}}\left(\frac{\partial\tilde{H}_y}{\partial x} - \frac{\partial\tilde{H}_x}{\partial y}\right) = \hat{\mathbf{z}}j\omega\epsilon\tilde{E}_z$$

$$\frac{\partial\tilde{H}_y}{\partial x} = \frac{\partial\tilde{H}_x}{\partial y} = 0, \quad \therefore \tilde{E}_z = 0$$

A plane wave has no electric- or magnetic-field components along its direction of propagation

Uniform Plane Waves (cont'ed)



$$\begin{aligned} \tilde{E}_x(z) &= \tilde{E}_x^+(z) + \tilde{E}_x^-(z) \\ &= E_{x0}^+ e^{-jkz} + E_{x0}^- e^{jkz} \end{aligned}$$

$+z$ -direction $-z$ -direction

Uniform Plane Waves (cont'ed)

Objective: To get the relationship of \mathbf{E} and \mathbf{H} in plane wave propagation

Assume \mathbf{E} has only a component along x -direction, $E_y=0$

$$\nabla \times \mathbf{E} = \begin{vmatrix} \hat{\mathbf{x}} & \hat{\mathbf{y}} & \hat{\mathbf{z}} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ \tilde{E}_x^+(z) & 0 & 0 \end{vmatrix}$$

$$= -j\omega\mu(\hat{\mathbf{x}}\tilde{H}_x + \hat{\mathbf{y}}\tilde{H}_y + \hat{\mathbf{z}}\tilde{H}_z)$$



$$\tilde{H}_x = \tilde{H}_z = 0$$

$$\tilde{H}_y = \frac{1}{-j\omega\mu} \frac{\partial \tilde{E}_x^+(z)}{\partial z}$$

$V_0^+ = Z_0 I_0^+$



$$\tilde{H}_y(z) = \frac{k}{\omega\mu} E_{x0}^+ e^{-jkz} = H_{y0}^+ e^{-jkz}$$

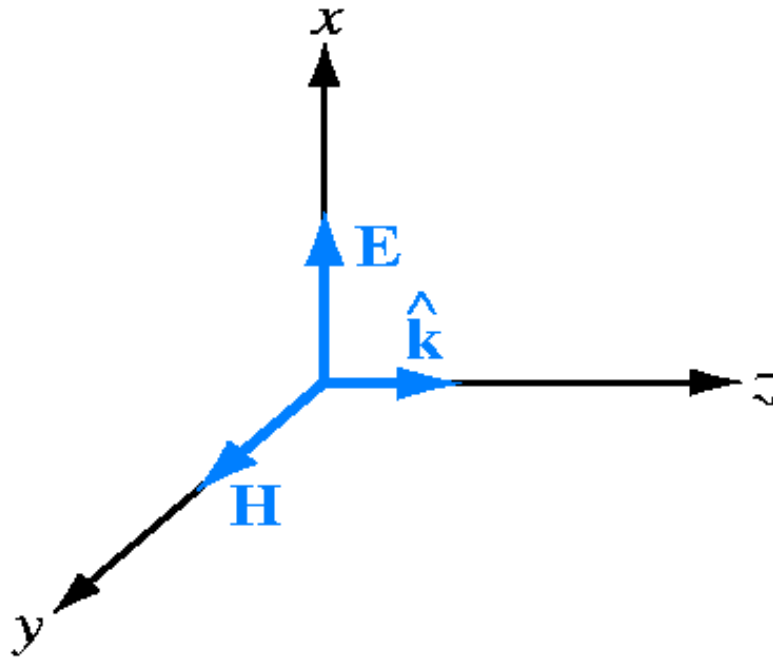


$$H_{y0}^+ = \frac{k}{\omega\mu} E_{x0}^+$$

Intrinsic impedance

$$\eta = \frac{\omega\mu}{k} = \sqrt{\frac{\mu}{\epsilon}}$$

Transverse electromagnetic (TEM) wave



$$\begin{aligned}\tilde{\mathbf{E}}(z) &= \hat{\mathbf{x}}\tilde{E}_x^+(z) \\ \tilde{\mathbf{H}}(z) &= \hat{\mathbf{y}}\frac{\tilde{E}_x^+(z)}{\eta}\end{aligned}$$

complex

$$\tilde{E}_x^+(z) = E_{x0}^+ e^{-jkz} = |E_{x0}^+| e^{j\phi^+} e^{-jkz}$$

Transverse electromagnetic (TEM) wave...

The instantaneous electric and magnetic fields are given by:

$$\begin{aligned}\mathbf{E}(z, t) &= \text{Re}[\tilde{\mathbf{E}}(z)e^{j\omega t}] \\ &= \hat{\mathbf{x}}|E_{x0}^+|\cos(\omega t - kz + \phi^+) \quad (\text{V/m}) \\ \mathbf{H}(z, t) &= \text{Re}[\tilde{\mathbf{H}}(z)e^{j\omega t}] \\ &= \hat{\mathbf{y}}\left|\frac{E_{x0}^+}{\eta}\right|\cos(\omega t - kz + \phi^+) \quad (\text{A/m})\end{aligned}$$

They are *in phase*

Phase velocity for Plane Wave Propagation

Recall:

$$y(x, t) = A \cos\left(2\pi f t - \frac{2\pi}{\lambda} x\right) \\ = A \cos(\omega t - \beta x)$$

$$u_p = \frac{\omega}{\beta} = f\lambda$$

$$\beta = \frac{2\pi}{\lambda} \quad (\text{rad/m})$$

Deduce from:

$$\mathbf{E}(z, t) = \text{Re}[\tilde{\mathbf{E}}(z)e^{j\omega t}] \\ = \hat{\mathbf{x}} |E_{x0}^+| \cos(\omega t - kz + \phi^+) \quad (\text{V/m})$$

$$u_p = \frac{\omega}{k} = \frac{\omega}{\omega\sqrt{\mu\varepsilon}} = \frac{1}{\sqrt{\mu\varepsilon}} \quad (\text{m/s})$$

$$\lambda = \frac{2\pi}{k} = \frac{u_p}{f} \quad (\text{m})$$



If the medium is vacuum, $u_p = ?$ $\eta = ?$

Example 6-5: EM Plane Wave in Air

The electric field of a 1-MHz plane wave traveling in the $+z$ -direction in air points along the x -direction. If the peak value of E is 1.2π (mV/m) and E is max. at $t=0$ and $z=50$ m, obtain expressions for $\mathbf{E}(z,t)$ and $\mathbf{H}(z,t)$, then plot these variations as a function of z at $t=0$.

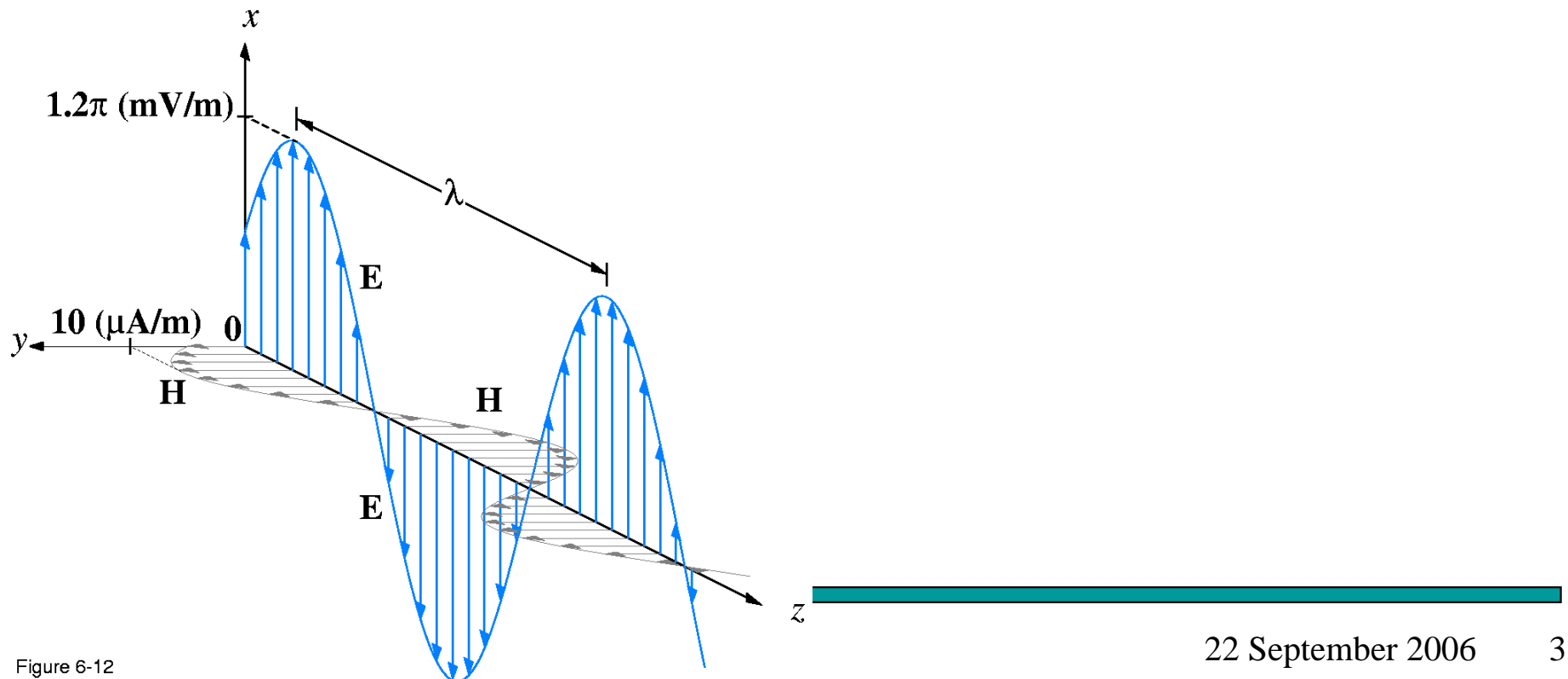
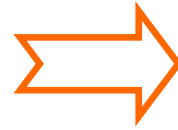
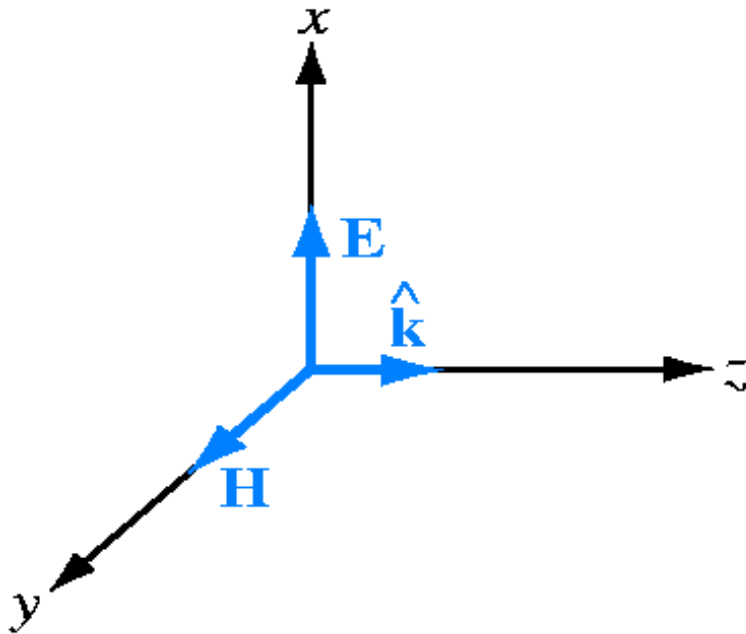


Figure 6-12

General Relation between **E** and **H**



$$\tilde{\mathbf{H}} = \frac{1}{\eta} \hat{\mathbf{k}} \times \tilde{\mathbf{E}}$$

$$\tilde{\mathbf{E}} = -\eta \hat{\mathbf{k}} \times \tilde{\mathbf{H}}$$

*When we rotate the four fingers of the right hand from the direction of **E** toward the direction of **H**, the thumb will point in the direction of wave travel, **k-hat***

Exercise

A 10-MHz uniform plane wave is traveling in a nonmagnetic medium with $\mu = \mu_0$ and $\epsilon_r = 9$. Find

- (a) The phase velocity
- (b) The wavenumber
- (c) The wavelength in the medium
- (d) The intrinsic impedance of the medium

Thank You