

Multimode network equivalence of waveguide discontinuities using full-wave method of moments for spatial power combining

Rakesh Singh Kshetrimayum and Lei Zhu

*School of Electrical and Electronic Engineering
Nanyang Technological University, SINGAPORE*

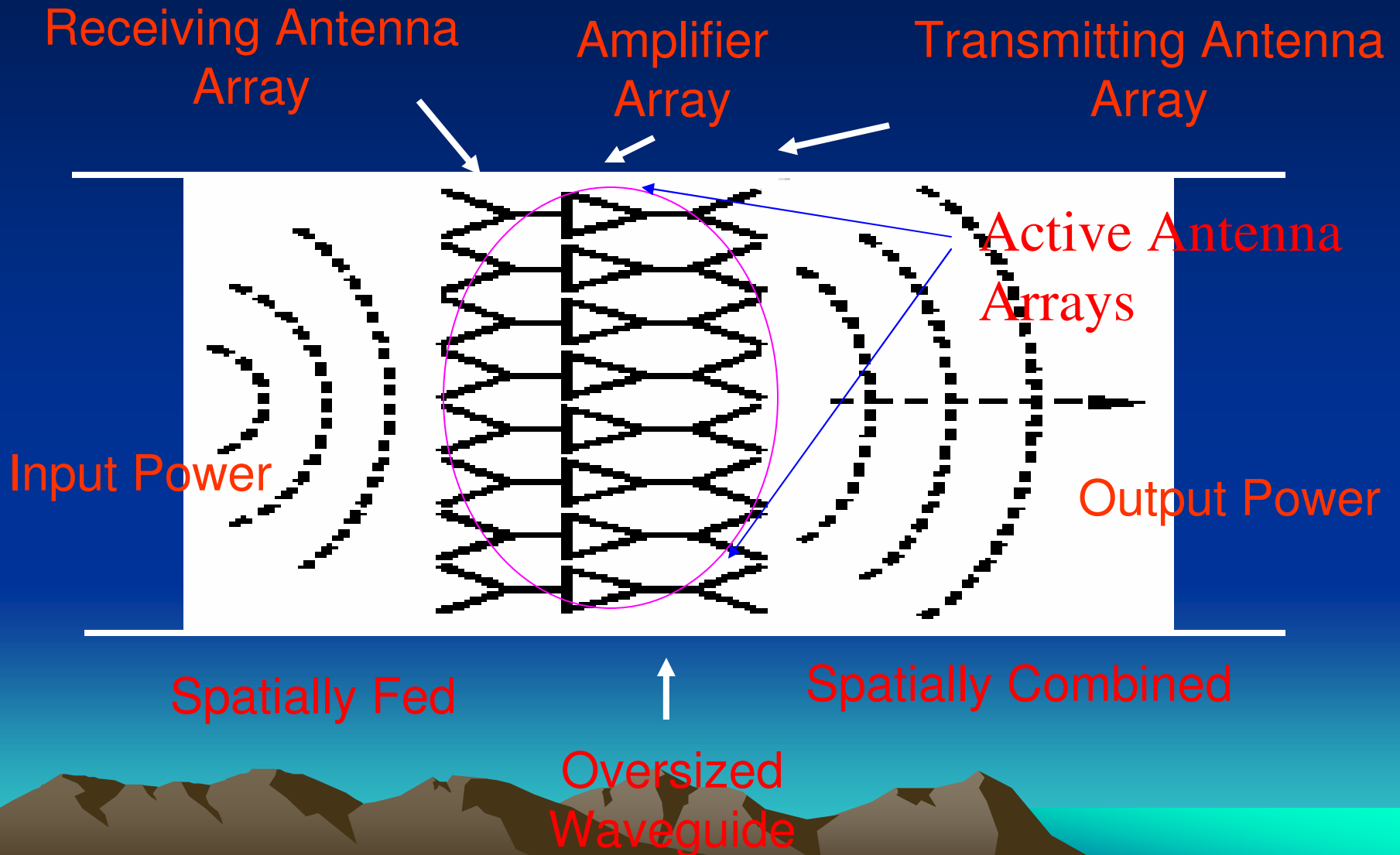


Scope of the presentation

- *Introduction to Spatial Power Combining*
- *Development of full-wave method of moments for electric-type (patch, strip) antenna inside oversized waveguide*
- *Multimode network equivalence of oversized waveguide discontinuities*



Waveguide-based spatial power combining

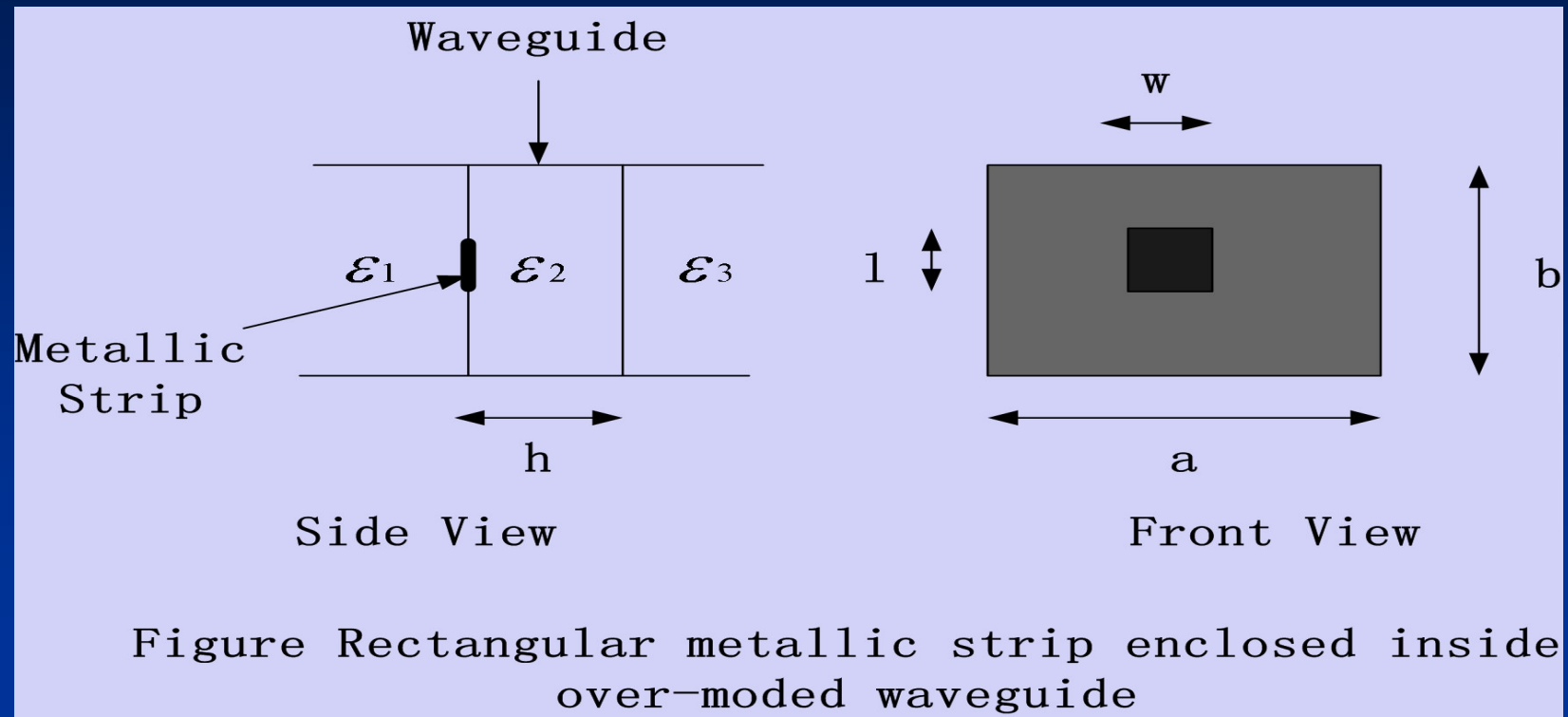


Modeling Issues

- *Oversized waveguide to accommodate large active antenna arrays*
- *Oversized waveguides: several high-order propagating waveguide modes in addition to the TE₁₀ dominant mode.*
- *But commercial softwares can only model waveguide structures for TE₁₀ mode only.*



Modeling Issues (Contd.)



- Hence it is necessary to develop a program which can characterize the enclosed metallic strip for multimode case.*

Development of full-wave method of moments

- *S.-G. Pan and I. Wolff, “Scalarization of Dyadic Green’s functions and network formalism for three-dimensional full-wave analysis of planar lines and antennas,” in IEEE Trans. MTT, Nov. 1994*



Spectral Dyadic Green's functions

- The spectral dyadic Green's functions for the layered waveguide is given below.*

$$\hat{G}_{Ei}(k_i, z, z') = \frac{jZ_1''Z_2'' \tan \gamma_2 h (\hat{k}_i \times z)(\hat{k}_i \times z)}{Z_1'' + jZ_2'' \tan \gamma_2 h} + \frac{jZ_1'Z_2' \tan \gamma_2 h (\hat{k})(\hat{k})}{Z_1' + jZ_2' \tan \gamma_2 h}$$

$$Z(z) = \frac{1}{Y(z)} = \begin{cases} Z^e = \frac{k_z(z)}{\omega \epsilon_i(z)} \\ Z^h = \frac{\omega \mu_i(z)}{k_z(z)} \end{cases} \quad k_z^2(z) = \omega^2 \mu_i \epsilon_i - k_i^2$$

Space domain dyadic Green's functions

The spectral dyadic Green's functions can be transformed into their space domain counterparts:

$$G_{ij}(i, j; i', j') = \sum_{m=0}^{M+1} \sum_{n=0}^{N+1} \tilde{G}_{ij}(k_x, k_y) \phi_{mn}^i \phi_{mn}^j$$

$$\phi_{mn}^x = \sqrt{\frac{\varepsilon_{0m}}{a}} \sqrt{\frac{\varepsilon_{0n}}{b}} \cos(k_x x) \cos(k_y y)$$

$$\phi_{mn}^y = \sqrt{\frac{\varepsilon_{0m}}{a}} \sqrt{\frac{\varepsilon_{0n}}{b}} \sin(k_x x) \cos(k_y y)$$

ε_{0m} and ε_{0n} are Neuman indices where $\varepsilon_{0m} = 1$ for $m=0$ and $\varepsilon_{0m} = 2$ for $m \neq 0$.

Electric Field Integral Equation (EFIE)

$$\hat{z} \times E(r) + j\omega\mu \hat{z} \times \left(\int_{strip} \hat{G}_{EJ}(r, r') \cdot J(r') dS' \right) = 0$$

The vector EFIE can be reduced to a coupled set of scalar integral equations as below:

$$E_x(x, y) = j\omega\mu \int_{strip} \int G_{xx}(x, y; x', y') \cdot J_x(x', y') dx' dy' + \int_{strip} \int G_{xy}(x, y; x', y') \cdot J_y(x', y') dx' dy'$$

$$E_y(x, y) = j\omega\mu \int_{strip} \int G_{yx}(x, y; x', y') \cdot J_x(x', y') dx' dy' + \int_{strip} \int G_{yy}(x, y; x', y') \cdot J_y(x', y') dx' dy'$$



Galerkin's Method

- *Galerkin's method is applied to the EFIE to transform the integral equations into matrix systems of linear equations*
- *Piece-wise sinusoidal and pulse functions are used as basis and testing functions*
- $[Z][I]=[V]$

where

$$[Z] = \begin{bmatrix} Z_{xx} & Z_{xy} \\ Z_{yx} & Z_{yy} \end{bmatrix} \quad [I] = \begin{bmatrix} I_x \\ I_y \end{bmatrix} \quad [V] = \begin{bmatrix} V_x \\ V_y \end{bmatrix}$$

Matrix systems of linear equations

where the matrix components are given by

$$Z_{ij} = j\omega\mu \sum_{m=0}^{M+1} \sum_{n=0}^{N+1} \tilde{j}_i(k_x, k_y) \tilde{G}_{ij}(k_x, k_y) \tilde{j}_j(k_x, k_y)$$

$$V_x = -\int \int_{\text{strip}} J_x(x, y) E_x(x, y) dx dy$$

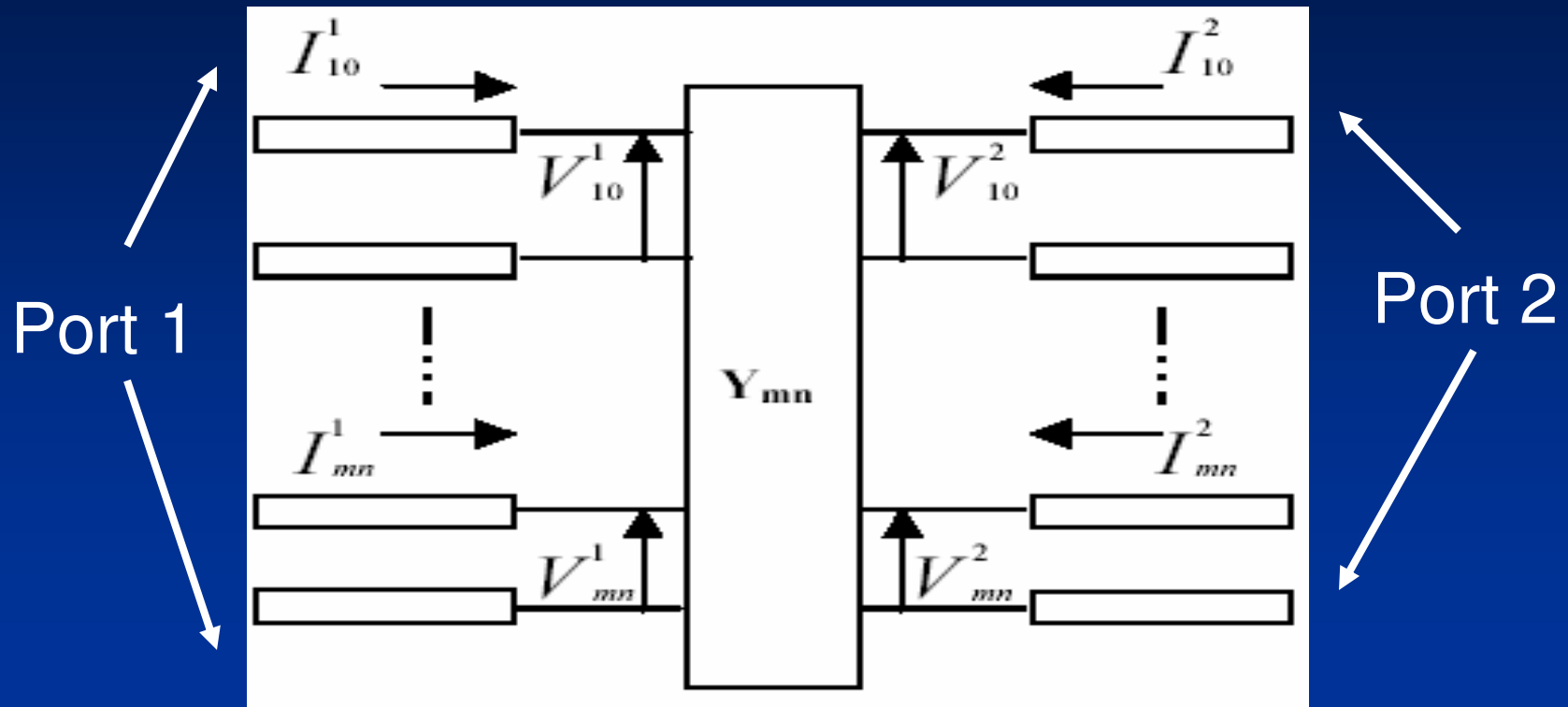
I_x and I_y are the unknown current coefficients



Electromagnetic program

- *An EM program is developed to characterize enclosed metallic strips*
- *Initial results obtained for waveguide dimension $a=0.02286$, $b=0.01016$, strip dimension $w=0.009271$, $l=0.007112$.*
- *Strip is placed at the centre of the over-moded waveguide. substrate of $\epsilon_2=2.33$, layer thickness $h=0.00136$. $\epsilon_1=1.0$ $\epsilon_3=1.0$.*

Multimode Network Equivalent



Multimode Network Equivalence for metallic strip enclosed inside oversized waveguide

Thank you

Have a Nice Day!

